

Lecture 11

10/11/12

Linear Systems

last time:

Modal representation of responses of (Unforced) LTI systems

Fact: If A is diagonalizable then the solution of

$$\dot{x}(t) = Ax(t), \quad x(0) = x_0$$

can be written as:

$$\begin{bmatrix} x(t) \end{bmatrix} = \sum_{i=1}^n e^{\lambda_i t} \begin{bmatrix} v_i \end{bmatrix} \begin{bmatrix} w_i^* \end{bmatrix} \begin{bmatrix} x_0 \end{bmatrix}$$

Note! $Av_i = \lambda_i v_i$; v_i : right e-vectors (e-vectors)

$w_i^* A = \lambda_i w_i^*$; w_i : left e-vectors

$$(A^* w_i = \bar{\lambda}_i w_i)$$

Fact: w_i 's can always be selected st.

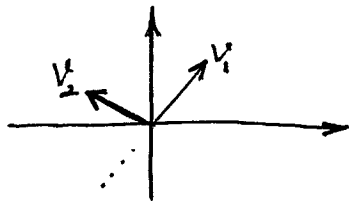
$$w_i^* v_j = \delta_{ij} = \begin{cases} 1 & i=j \\ 0 & i \neq j \end{cases}$$

$$Q: x_0 = v_m \Rightarrow x(t) = e^{\lambda_m t} \cdot v_m$$

If matrix A is normal $A^T A = A A^T$

then v_i 's are mutually orthogonal and $w_i = v_i$

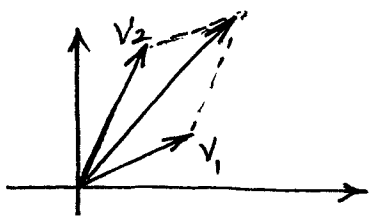
e.g. for $n=2$



$$\begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = c_1 e^{\lambda_1 t} \begin{bmatrix} v_1 \end{bmatrix} + c_2 e^{\lambda_2 t} \begin{bmatrix} v_2 \end{bmatrix}$$

$\downarrow$
 $w_2^* x_0$

For non-normal matrices: we can have e-vectors that are far from being orthogonal (nearly parallel)



e.g. $A = \begin{bmatrix} -1 & 0 \\ k & -2 \end{bmatrix}$

$k=0$ A diagonal

as k grows A becomes more

ill conditioned and e-vectors are going to move from orthogonal to almost parallel.

For systems with inputs :

$$x(t) = e^{At} x_0 + \int_0^t e^{A(t-\tau)} \underbrace{B \cdot u(\tau)}_{q(\tau)} d\tau$$

$$= \sum_{i=1}^n e^{\lambda_i t} v_i w_i^* x_0 + \sum_{i=1}^n \int_0^t e^{\lambda_i(t-\tau)} v_i w_i^* q(\tau) d\tau$$

Recall : if matrices are diagonalizable we can have

$$\dot{x}(t) = A x(t)$$

can be brought into the following form: $\dot{z}(t) = \Lambda z(t)$

where $A = V \Lambda V^{-1}$

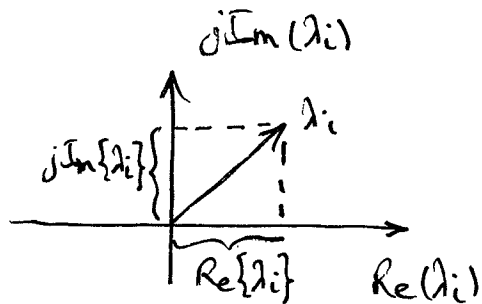
$$\Lambda = \begin{bmatrix} \lambda_1 & & \\ & \lambda_2 & \\ & & \ddots \\ & & & \lambda_n \end{bmatrix} \Rightarrow z_i(t) = e^{\lambda_i t} z_i(0) ; t \in [0, \infty)$$

$i = 1, \dots, n$

In D.T.

$$z_i(t) = \lambda_i^T z_i(0) ; t = 0, 1, 2, \dots$$

$$\lambda_i \in \mathbb{C}$$



In C.T.

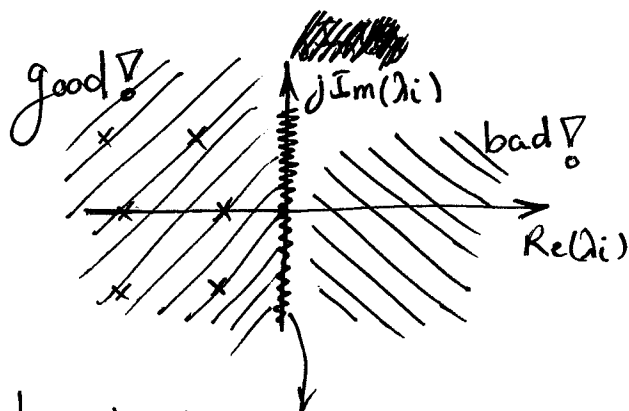
$$Z_i(t) = e^{(\text{Re}(\lambda_i) + j\text{Im}(\lambda_i))t} \cdot Z_i(0)$$

$$Z_i(t) = e^{\text{Re}(\lambda_i)t} e^{j\text{Im}(\lambda_i)t} \cdot Z_i(0)$$

Note! $\text{Re}(\lambda_i)$ determines growth or decay rate with time

$$\text{Re}(\lambda_i) < 0 \Rightarrow \text{decay}$$

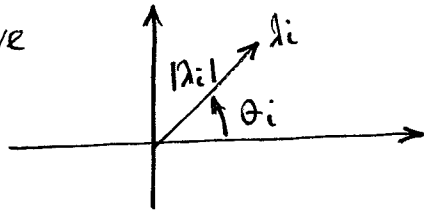
$$\text{Re}(\lambda_i) > 0 \Rightarrow \text{grow}$$



Careful! Imaginary axis deserves more attention!

In DT: $z_i(t) = \lambda_i^t z_i(0)$

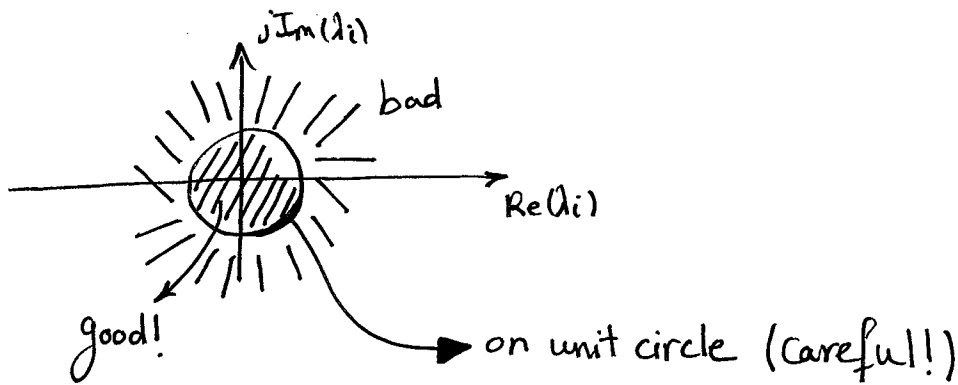
we have



In polar coordinates we have:

$$z_i(t) = (|\lambda_i| e^{j\theta_i})^t z_i(0)$$

$$z_i(t) = |\lambda_i|^t e^{j\theta_i t} \cdot z_i(0)$$



Recall: Double integrator

$$A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$$

$$\lambda_1 = \lambda_2 = 0$$

$$e^{At} = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix}$$

In general for Jordan blocks:

$$J_3 = \begin{bmatrix} \lambda & 1 & 0 \\ 0 & \lambda & 1 \\ 0 & 0 & \lambda \end{bmatrix} \Rightarrow e^{J_3 t} = e^{\lambda t} = \begin{bmatrix} 1 & t & \frac{1}{2}t^2 \\ 0 & 1 & t \\ 0 & 0 & 1 \end{bmatrix}$$

$\text{Re}(\lambda) < 0 \Rightarrow$ asymptotic decay (as $t \rightarrow \infty$)

If $\text{Re}(\lambda) = 0 \Rightarrow$ response will blow up as $t \rightarrow \infty$.

Summary of conditions for stability of LTI systems
(unforced)
asymptotic decay in time

$$\dot{x}(t) = Ax(t) \quad \dots \quad (*)$$

(In terms of e-values of A)

System $(*)$ is 1. stable $\Leftrightarrow \text{Re}(\lambda_i) < 0$
 $i = 1, \dots, n$

2. Unstable if either of these hold:

(a) There is λ_i st. $\text{Re}(\lambda_i) > 0$

(b) There is λ_i st. $\text{Re}(\lambda_i) = 0$ with algebraic multiplicity greater than or equal to 2.

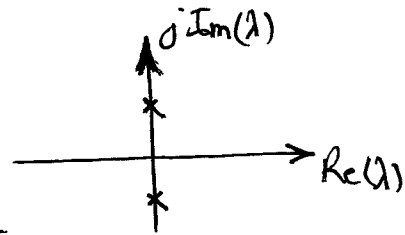
Ex

$$A = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$

$$\lambda_1 = j$$

$$\lambda_2 = -j$$

algebraic
multiplicity 1.



3. Marginally stable if all e-values have $\text{Re}(\lambda_i) \leq 0 \quad i=1, \dots, n$ and the e-values with $\text{Re}(\lambda_i) = 0$ have algebraic multiplicity one (no repeated e-values on $j\omega$ -axis)

Ex $A = \begin{bmatrix} 0 & 1 \\ -a_0 & -a_1 \end{bmatrix}$

$$\det(sI - A) = s^2 + a_1 s + a_0$$

1) stable $\Leftrightarrow \begin{matrix} a_1 > 0 \\ a_0 > 0 \end{matrix}$ (Routh-Hurwitz criterion)

2) $a_1 = 0$

$$\det(sI - A) = s^2 + a_0$$

$$a_0 < 0 \Rightarrow \det(sI - A) = s^2 - |a_0| \Rightarrow s_1 = \sqrt{|a_0|}$$

$$s_2 = -\sqrt{|a_0|}$$

2(a) applying $\Rightarrow$ unstable

inverted pendulum

$$a_0 = 0 \Rightarrow \det(sI - A) = s^2$$

$$s_1 = 0 ; s_2 = 0 \Rightarrow \text{from 2(b)} \rightarrow \text{unstable}$$

↳ double integrator

$$a_0 > 0 \Rightarrow \begin{aligned} s_1 &= +j\sqrt{a_0} \\ s_2 &= -j\sqrt{a_0} \end{aligned}$$

⇒ case (3) Marginally stable

↓
mass spring system

$$3) a_0 = 0$$

$$\det(sI - A) = s^2 + a_1 s$$

$$\left. \begin{aligned} s_1 &= 0 \\ s_2 &= -a_1 \end{aligned} \right\} \begin{aligned} &\text{Marginally stable if } a_1 > 0 \\ &\text{unstable if } a_1 \leq 0 \end{aligned}$$