

Lecture 24

Linear Systems

12,04,12

Last time

Many things ... (Crb, obsv, ...)

Today _____

- pole placement
-
- If time permits :
- observer-based design
(output feedback)

Fact: If system $\dot{x} = Ax + Bu$ (1)

is ctrb. then we can design a state-feedback

controller :

$$u(t) = -Kx(t) \quad (2)$$

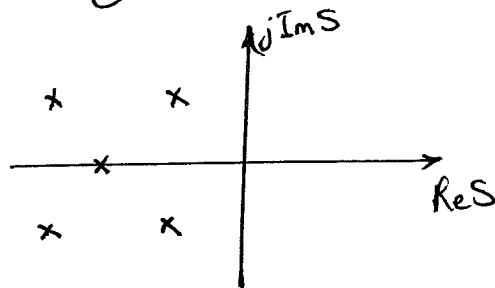
control signal at time t feedback gain matrix state measurement at time t

[Matrix of feedback gains]

to place e-values of a closed-loop A-matrix :

$$A_d = A - BK$$

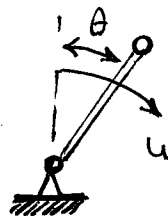
to an arbitrary location in the LHP



(2) $\rightarrow$ (1) closed-loop system.

$$\dot{x} = Ax + B(-Kx) \Rightarrow \dot{x} = \underbrace{(A - BK)}_{A_d} x$$

Ex.



$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$u(t) = -Kx(t) = -[k_1 \ k_2] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

$$k_1, k_2 \in \mathbb{R}$$

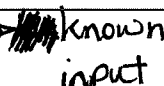
$$A_d = \begin{bmatrix} 0 & 1 \\ 1 - k_1 & -k_2 \end{bmatrix}$$

$$f(s) = \det(sI - A) = s^2 + K_2 s + K_1 - 1$$

$$= (s - \lambda_1)(s - \lambda_2) = s^2 - (\lambda_1 + \lambda_2)s + \lambda_1 \lambda_2$$

(Matlab) use the command: `place(A, B, [\lambda_1, \lambda_2])`
which would K as an output.

* Observer design (Estimator)

state eq: $\dot{x} = Ax + Bu$  known input

measured output: $y = Cx$ 

Objective: Design a dynamical system [model that we
(observer, estimator) want to simulate]

s.t.

$$\hat{x}(t) \xrightarrow{t \rightarrow \infty} x(t)$$

 state of estimator (observer)

The problem from simulating this system comes from
not knowing the initial conditions of the system $x(0)$

If we knew $x(0)$, we could simulate $\textcircled{*}$ but we don't
(know $x(0)$)

An attempt (at designing an observer) "copy of $\textcircled{*}$ "

$$\begin{aligned} \dot{\hat{x}}(t) &= A \hat{x}(t) + Bu & ; \hat{x}(0) : \text{under our control} \\ \hat{y}(t) &= C \hat{x}(t) \end{aligned} \quad \left. \vphantom{\begin{aligned} \dot{\hat{x}}(t) &= A \hat{x}(t) + Bu \\ \hat{y}(t) &= C \hat{x}(t) \end{aligned}} \right\} \begin{array}{c} \text{Naive} \\ \uparrow \\ \textcircled{NO} \end{array}$$

If we simulate $\textcircled{NO}$ will we get $\hat{x}(t) \xrightarrow{t \rightarrow \infty} x(t)$?

Form: $\tilde{x}(t) = x(t) - \hat{x}(t)$ and check if $\tilde{x}(t) \xrightarrow{t \rightarrow \infty} 0$.

$\downarrow$
 estimation error

$\textcircled{*} - \textcircled{NO}$:

$$\dot{x} - \dot{\hat{x}} = A \cdot (x - \hat{x})$$

$$y - \hat{y} = C(x - \hat{x})$$

$$\dot{\tilde{x}} = A \tilde{x}(t)$$

$$\tilde{y} = C \tilde{x}(t)$$

Q: $\tilde{x}(t) \xrightarrow{t \rightarrow \infty} 0$

A: Yes if $\lambda_i(A) \in \text{LHP } i=1, \dots, n$

$$\tilde{x}(0) = x(0) - \hat{x}(0) \neq 0$$

? $\nwarrow$ $\nearrow$ known

Ex I.P.

$$y = x_1$$

$$x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} \text{known} \\ ? \end{bmatrix}$$

$$y = x_1$$

We will have problems for unstable systems
[e.g. inverted pendulum] because of e-values that
are not in LHP.

Fix

(**)
$$\begin{cases} \dot{\hat{x}}(t) = A\hat{x}(t) + Bu(t) + L(y(t) - \hat{y}(t)) \\ \hat{y} = C\hat{x}(t) \end{cases}$$

a "copy" of our system

observer gain matrix L

measured output $y(t)$

estimated output $\hat{y}(t)$

an injection term $(y(t) - \hat{y}(t))$

(**) :
$$\dot{\hat{x}} = A\hat{x} + Bu + LC(x - \hat{x}) \quad \dots (I)$$

from (*) - (I) we get:

$$\dot{x} - \dot{\hat{x}} = A(x - \hat{x}) + \cancel{Bu} - \cancel{Bu} - LC(x - \hat{x})$$

$$y - \hat{y} = C(x - \hat{x})$$

$$\Rightarrow \quad \begin{aligned} \dot{\tilde{x}} &= (A-LC) \tilde{x} \\ \tilde{y} &= C \tilde{x} \end{aligned}$$

Conclusion: If $A-LC$ is a stable matrix

$$\tilde{x}(t) \xrightarrow{t \rightarrow \infty} 0$$

(our ability to drive $\hat{x}(t)$ to $x(t)$ as $t \rightarrow \infty$ depends on whether we could choose L st. $\lambda_i(A-LC) \in \text{LHP}$)

Ex. I.P.

$$A-LC = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} l_1 \\ l_2 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} = \begin{bmatrix} -l_1 & 1 \\ 1-l_2 & 0 \end{bmatrix}$$

V.S.

$$\begin{aligned} A-BK &= \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} k_1 & k_2 \end{bmatrix} \\ &= \begin{bmatrix} 0 & 1 \\ 1-k_1 & -k_2 \end{bmatrix} \end{aligned}$$

Fact: e-values of $A-LC \equiv$ e-values of $\underbrace{(A^T)}_A - \underbrace{(C^T)}_{\bar{B}} \underbrace{(L^T)}_{\bar{K}} = (A-LC)^T$

e-values don't change by transposing a matrix.

Thus: our ability to move e-values of A-LC to LHP is determined by controllability of pair (A^T, C^T)



observability of "pair" (A, C)

$$\text{rank} [C^T \mid A^T C^T \mid \dots \mid (A^{n-1})^T C^T] = n$$

$$\updownarrow$$

$$\text{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} = n \iff \begin{array}{l} \dot{x} = Ax \\ y = Cx \end{array} \text{ observable!}$$
