

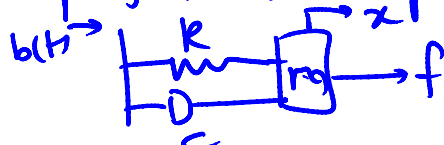
Lecture 16

Thursday, March 24, 2011
8:14 AM

⊛ Systems are often described in terms of input-output transfer function or in terms of actual realizations

- input-output systems implicitly assumes an "input" signal and an output signal.
- A ~~phys~~ realized physical is usually abundant of ODE, PDEs.

Example: Spring-mass-damper system



$$m\ddot{x} = f + k(b-x) - c\dot{x}$$

$$= m\ddot{x} + c\dot{x} + kx = f + kb(t)$$

$$y = x ;$$

$$y = \left[\frac{f(s)}{ms^2 + cs + k} + \frac{k b(s)}{ms^2 + cs + k} \right]$$

$$y = \left[\frac{1}{ms^2 + cs + k} \right] f(s) + \left[\frac{k}{ms^2 + cs + k} \right] b(s)$$

$$Y = \begin{bmatrix} x \\ \dot{x} \end{bmatrix} = \begin{bmatrix} \frac{f(s)}{ms^2 + cs + k} \\ \frac{s f(s)}{ms^2 + cs + k} \end{bmatrix} \text{ with } f \text{ as the input}$$

⊛ a) Given a transfer function, what is

⊙ ⊙ Given a transfer function what is the "minimal" physical realization of the transfer function?

State-space approach :

$$\ddot{p} + \dot{p} + kp = f \quad ; \quad x = \begin{bmatrix} p \\ \dot{p} \end{bmatrix}$$

$$\underline{\underline{\dot{x} = Ax + Bf}} \quad ; \quad A = \begin{bmatrix} 0 & 1 \\ -k & -1 \end{bmatrix} ; B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

↑
inputs.

Here the notion of stability is asymptotic stability: when does the solution $x(t) \rightarrow 0$ as $t \rightarrow \infty$ for any initial condition $x(0)$ with the input set to zero.

$$x(t) = e^{At} x(0) + \int_0^t e^{A(t-z)} B u(z) dz$$

(Variation of parameters formula)

$$e^{At} = I + At + \frac{A^2 t^2}{2!} + \frac{A^3 t^3}{3!} + \dots$$

$A \in \mathbb{R}^{n \times n}$

The initial condition response

$$x(t) = e^{At} x(0).$$

For a.s. $x(t) \rightarrow 0$ as $t \rightarrow \infty$ for any $x(0)$

$\Leftrightarrow \lambda_i(A)$ is in the LHP for eigenvalues λ_i of A .

$$\begin{aligned} \dot{x} &= Ax + Bu & ; & \quad a.s \Leftrightarrow \lambda_i(A) \in \text{LHP} \\ y &= Cx + Du & ; & \quad y \text{ is the output equation.} \end{aligned}$$

↓ output ↑ input

Let's assume initial conditions are zero
 $x(0) = 0$

$$sX(s) - x(0) = AX(s) + BU(s)$$

$$\Rightarrow (sI - A)X(s) = x(0) + BU(s)$$

$$\Rightarrow X(s) = (sI - A)^{-1}x(0) + (sI - A)^{-1}BU(s)$$

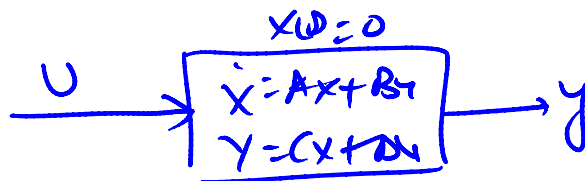
$x(0) = 0$

$$X(s) = (sI - A)^{-1}BU(s)$$

$$\begin{aligned} Y(s) &= CX(s) + DU(s) \\ &= C(sI - A)^{-1}BU(s) + DU(s) \end{aligned}$$

$$\therefore Y(s) = [C(sI - A)^{-1}B + D]U(s)$$

Summary: Given a physical realization and identified inputs and outputs we have obtained a transfer function relating input and the output.



$$U \rightarrow \boxed{C(sI - A)^{-1}B + D} \rightarrow Y$$



The input-output transfer function matrix is stable iff all poles of all elements are in the strict lhp.

→ When are the two notions of stability equivalent?

→ If A is a.s then any input-output transfer function will be stable

Pf: $Y = (C(sI-A)^{-1}B + D)U$

$$G = \frac{C \text{Adj}[(sI-A)]B + D}{\det(sI-A)}; \quad p_i = \frac{\text{Adj}(p)}{\det(p)}$$

$$= \frac{\{C \text{Adj}[(sI-A)]B + D \det(sI-A)\}}{\det(sI-A)} = \frac{n(s)}{d(s)}$$

where $n = C \text{Adj}[(sI-A)]B + D \det(sI-A)$

$$d = \det(sI-A)$$

If A has eigenvalues $\lambda_1, \lambda_2, \dots, \lambda_n$

$$\det(sI-A) = (s-\lambda_1)(s-\lambda_2) \dots (s-\lambda_n)$$

$$G = \left(\frac{C \text{Adj}[(sI-A)]B + D \det(sI-A)}{(s-\lambda_1)(s-\lambda_2) \dots (s-\lambda_n)} \right)$$

It follows that as $\lambda_i \in \text{lhp}$, G has all poles in the lhp.

∴ a.s \Rightarrow i.o. stability

\therefore a.s. \Rightarrow i.o. stability

It is easily possible that $G(s)$ is stable in the t.o. sense but A is not stable

\rightarrow these involve \therefore

$CAd, (SI-A)B + D \det(SI-A)$
having common unstable factors with $\det(SI-A)$.

i.o. stability $\not\Rightarrow$ a.s.
(with given choice of input and output)

② When does $C(SI-A)^{-1}B + D$ stable as a transfer matrix $\Leftrightarrow A$ is a.s.?

Ans: Sol Answer: (C,A) has to be ~~as~~ detectable and (B,A) has to be stabilizable.

\rightarrow Observability:

$$x(t) = e^{At}x(0) + \left(\int_0^t e^{A(t-z)} B u(z) dz \right)$$

(Control input is known)

$$y(t) = C e^{At} x(0) + \int_0^t e^{A(t-z)} B u(z) dz + D u(t)$$

$$\underbrace{\left[y(t) - C \int_0^t e^{A(t-z)} B u(z) dz - D u(t) \right]}_{m(t)} = C e^{At} x(0)$$

$$m(t) = C e^{At} x(0)$$

$$m(t) = C e^{At} x(0)$$

$$\frac{dm(t)}{dt} = CAe^{At} x(0)$$

$$\frac{d^2 m(t)}{dt^2} = CA^2 e^{At} x(0)$$

$$\frac{d^{n-1} m(t)}{dt^{n-1}} = CA^{n-1} e^{At} x(0)$$

$$\frac{d^n m(t)}{dt^n} = CA^n e^{At} x(0)$$

$$\begin{bmatrix} m(t) \\ m'(t) \\ \vdots \\ m^{(n-1)}(t) \end{bmatrix} = \begin{bmatrix} C e^{At} \\ CA e^{At} \\ \vdots \\ CA^{n-1} e^{At} \end{bmatrix} x(0)$$

$$\begin{bmatrix} m(0) \\ m'(0) \\ \vdots \\ m^{(n-1)}(0) \end{bmatrix} = \underbrace{\begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix}}_{\Theta} x(0)$$

$$x(0) = \Theta^{-1} \begin{bmatrix} m(0) \\ \vdots \\ m^{(n-1)}(0) \end{bmatrix}$$

$$x(t) = e^{At} \Theta^{-1} \begin{bmatrix} m(0) \\ \vdots \\ m^{(n-1)}(0) \end{bmatrix} + \int_0^t e^{A(t-\tau)} B u(\tau) d\tau$$

Cayley Hamilton theorem gives the necessary

$$\dot{x} = Ax + Bu, \quad u = Kx + v$$

$$\underline{\dot{x} = Ax + BKx + Bv}$$

$$\dot{x} = (A + B_1 C_1)x + B_1 y$$

Controllability

$$x(t) = e^{At} x(0) + \int_0^t e^{A(t-z)} B u(z) dz$$

$$x_d = x(t_f)$$

$$x(t) = x_d$$

$$\int_0^{t_f} e^{A(t_f-z)} B u(z) dz = x_d - e^{A t_f} x(0)$$

$$\int_0^{t_f} e^{A(t_f-z)} B u(z) dz$$

$$\int_0^{t_f} e^{A(t_f-z)} B u(z) dz = \int_0^{t_f} e^{A(t_f-z)} B u(z) dz$$

$(F F^{-1})$