

1 Distributed Data Compression

From the previous lectures we have learned that for two data sources X and Y , one can achieve rates at $R_1 + R_2 \geq H(X + Y)$, with two separated encoders which ignore the source correlation. However, if the two sources X and Y are correlated with each other, an optimal joint encoder can achieve compression rates at $H(X, Y)$, by applying the Slepian Wolf Theorem.

Now let us consider a similar problem, suppose there are two set of data X and Y , which both have n number of elements $(x_1, x_2, x_3, \dots, x_n)$ and $(y_1, y_2, y_3, \dots, y_n)$. These two set of data are correlated and not independent. The data sets are sent to two different encoders before being compressed together and being sent to a single joint decoder, as the relation shown in Figure 1.

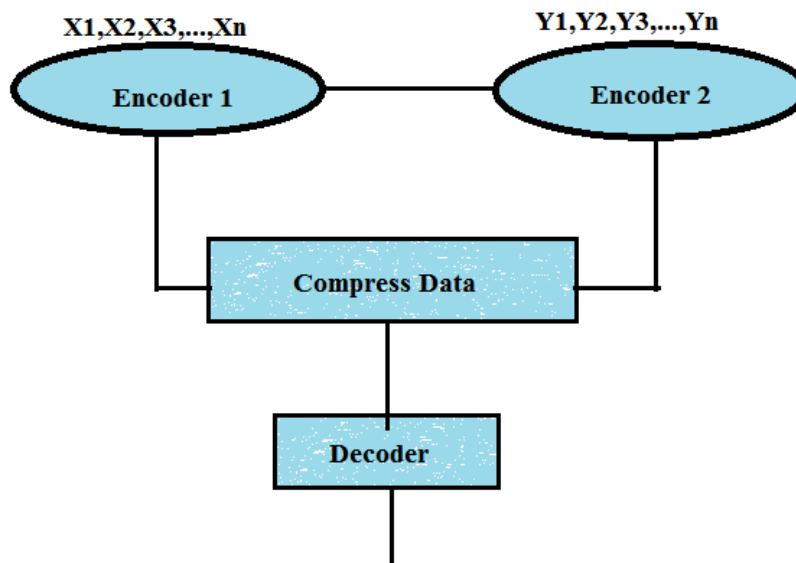


Figure 1: Block diagram of two correlated data sets

Such system is measured by the compress rate in bits per source symbol of the output streams of encoders. The single decoder is designed to be able to reconstruct the correlated data streams after compression in an optimal way. As we can see, encoder 1 has a rate of $R_1 = nH(X)$ while encoder 2 has a rate of $R_2 = nH(Y)$. The claim is if there is a communication channel between Encoder 1 and Encoder 2, by Slepian Wolf Theorem, the achievable rate of compression is:

$$nH(X) + nH(Y) \geq nH(X, Y) \quad \dots (1)$$

which means it can be compressed to $H(X, Y)$ bits/symbol.

1.1 Distributed Data Compression with Bernoulli Source

Now let us consider the following example: assume there are two correlated data sources $X(x_1, x_2, x_3 \dots x_n)$ and $Y(y_1, y_2, y_3 \dots y_n)$. X is Bernoulli(q) with the following properties:

$$\Pr(x_i = y_i) = 1 - p; \quad \Pr(x_i \neq y_i) = p.$$

Hence, Y is also Bernoulli($p + q + 2pq$) since:

$$\begin{aligned} \Pr(Y=1) &= \Pr(X = 1 \text{ and } Y = X) + \Pr(X = 0 \text{ and } Y \neq X) \\ &= q(1 - p) + (1 - q)p \\ &= p + q + 2pq \end{aligned}$$

The rate of compression we can achieve is to equal to: $nh(q) + nh(p + q + 2pq)$ bits. However, by the inequality (1) the optimal solution is to achieve $nh(q) + nh(p)$ bits.

Since:

$$\begin{aligned} n[H(X,Y)] &= n[H(X) + H(Y|X)] \\ &= n[H(X) + H(Z)] \end{aligned}$$

Where $Z_i = X_i + Y_i \text{ mod } 2$, for $i = 1$ to n

$$= n[h(q) + h(p)]$$

Notice it is always true that $p < \frac{1}{2}$, implies that:

$$q(1 - 2p) > 0 \quad \dots \dots \dots \gg \quad p + q + 2pq > p$$

and $nh(q) + nh(p)$ is an increasing function.

1.2 Distributed Data Compression over Mod-2 Arithmetic

Suppose there are two data sources $X(x_1, x_2, x_3 \dots x_n)$ and $Y(y_1, y_2, y_3 \dots y_n)$ over vector space $F^n = \{0, 1\}^n$. Let $Z = X + Y$. Notice that Z is also equal to $X - Y$ or $Y - X$, due to the fact that they are in mod-2 arithmetic.

Figure 2 shows the block diagram of sources X , Y and Z , where Z is Bernoulli(p).

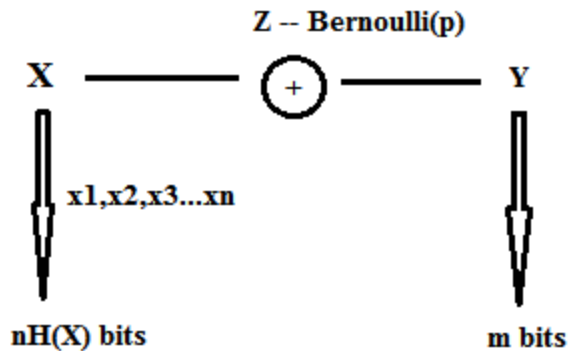


Figure 2

The rates of compression of X and Y is $nH(X)$ and m bits, respectively. From what was discussed above, the achievable number of bits per symbol for such problem is:

$$H(X) + \frac{m}{n} = h(q) + \frac{m}{n}$$

Then we claim that there exists an $m \times n$ ($m < n$) matrix Φ , such that $m \approx nh(p)$. With such matrix Φ , source data Y can be recovered from ΦY .

Here is the process of how to decode in the decoder: when we looking at the output of decoder, we have $nH(X)$ and ΦY . Then do the following:

- (1) decode X from $nH(X)$;
- (2) find ΦX simply by multiplying with each other;
- (3) find $\Phi Z = \Phi(X+Y) = \Phi X + \Phi Y$;
- (4) recover Z from ΦZ ;
- (5) find Y from $Y = X + Z$.

However, when recovering Z from ΦZ from step (4), there is no unique solution Z . But since any solution of Z would work for the problem purpose, we turn to the typical set of Z , with arbitrarily small error probability. First, let us put some definitions around here:

We know that Z is a Bernoulli(p), let $\Phi Z = b$ for simple. For the typical set of Z : $A_\epsilon^{(n)}(p)$ and $|A_\epsilon^{(n)}| \leq 2^{n(h(p)+\epsilon)}$. Now the goal is to find a vector \hat{Z} from the typical set $A_\epsilon^{(n)}$ that satisfies $\Phi \hat{Z} = b$. Such job could be complicated and require computer as a companion. But once we have the output \hat{Z} from whatever the tools we may used, we can recover Y from $Y = X + \hat{Z}$.

The probability of error $P_e^{(n)}$ can be calculated in bounds of:

$$\begin{aligned}
 P_e^{(n)} &\leq \Pr(Z \notin A_\epsilon^{(n)}(p)) + \Pr(\exists Z' \neq Z; Z \in A_\epsilon^{(n)}(p) \text{ and } \Phi Z' = b) \\
 &\leq \sum_{Z' \in A_\epsilon^{(n)}(p) \text{ and } Z' \neq Z} \Pr(\Phi Z' = b)
 \end{aligned}$$

To calculate $\Pr(\Phi Z' = b)$, we first choose Φ to be randomly uniformly with iid Bernoulli($\frac{1}{2}$) entries.

$$\begin{bmatrix} \Phi_{11} & \cdots & \Phi_{1n} \\ \vdots & \ddots & \vdots \\ \Phi_{m1} & \cdots & \Phi_{mn} \end{bmatrix} \times \begin{matrix} Z_1' \\ \vdots \\ Z_n' \end{matrix} = \begin{matrix} b_1 \\ \vdots \\ b_m \end{matrix}$$

Φ matrix: $m \times n$ Z' b

$$\Phi_{11}Z_1' + \Phi_{12}Z_2' + \cdots + \Phi_{1n}Z_n' = b_1$$

Since Φ is Bernoulli($1/2$), the probability of making Z' to equal b_i is equal to $\frac{1}{2^m}$, implies that:

$$\Pr(\Phi Z' = b) = \frac{1}{2} \times \frac{1}{2} \times \cdots \times \frac{1}{2} = \frac{1}{2^m}. \text{ Substitute it back the error inequality:}$$

$$\begin{aligned} P_e^{(n)} &\leq |A_\varepsilon^{(n)}(p)| \times \frac{1}{2^m} \\ &\leq 2^{n(h(p)+\varepsilon)} \times \frac{1}{2^m} \end{aligned}$$

$$P_e^{(n)} \text{ goes to } \frac{1}{2}n\varepsilon' \text{ if } m = n(h(p) + \varepsilon + \varepsilon').$$

Now that $nh(q) + m = n[h(q) + \frac{m}{n}] \approx n[h(q) + h(p)]$.

1.3 Distribution Data Compression when two sources are closed

Consider a similar problem with the case that the two sources $X(x_1, x_2, x_3 \dots x_n)$ and $Y(y_1, y_2, y_3 \dots y_n)$ are very closed to each other (ie. $X-Y$ to be minimum). The process diagram is shown in Figure 3 below:

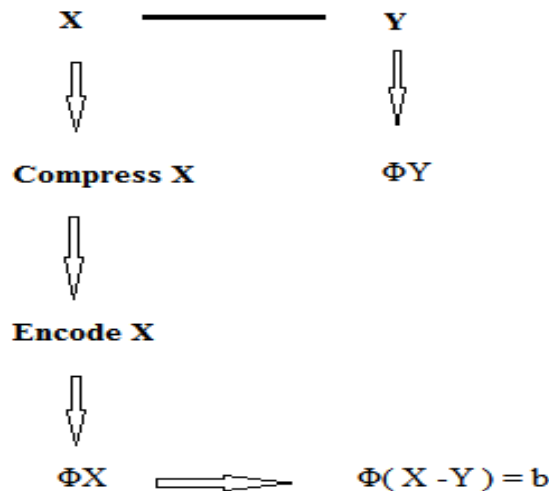


Figure 3

Let $Z = X - Y$, which can be considered as a noise function $N(0, \sigma^2)$. The job once again is to find Z from $\Phi Z = b$. Since the noise figure should be as small as possible, we want to find $\min \|Z\|_{l_2}^2$, such that $\Phi Z = b$.

♠ **Claim:** Assuming that $\Phi\Phi^T$ is nonsingular, $\hat{Z} = \Phi^T (\Phi\Phi^T)^{-1} b$

♠ **Proof:** suppose that Z satisfies $\Phi Z = b$.

$$\begin{aligned} \text{Then } \|Z\|_{l_2}^2 &= \|Z - \hat{Z} + \hat{Z}\|_{l_2}^2 \\ &= \|Z - \hat{Z}\|_{l_2}^2 + Z \hat{Z}^T (Z - \hat{Z}) + \|Z\|_{l_2}^2 \end{aligned}$$

The term of $Z \hat{Z}^T (Z - \hat{Z})$ goes to zero due to:

$$\begin{aligned} \hat{Z}^T (Z - \hat{Z}) &= b^T (\Phi\Phi^T)^{-1} \Phi (Z - \hat{Z}) \\ &= b^T (\Phi\Phi^T)^{-1} (b - b) \\ &= 0 \end{aligned}$$

Hence, $\|Z\|_{l_2}^2 \geq \|\hat{Z}\|_{l_2}^2$

Now we have: $\Phi \hat{Z} = \Phi\Phi^T (\Phi\Phi^T)^{-1} b$, which could be used as a solution.

2 Linear Error-Correcting Code

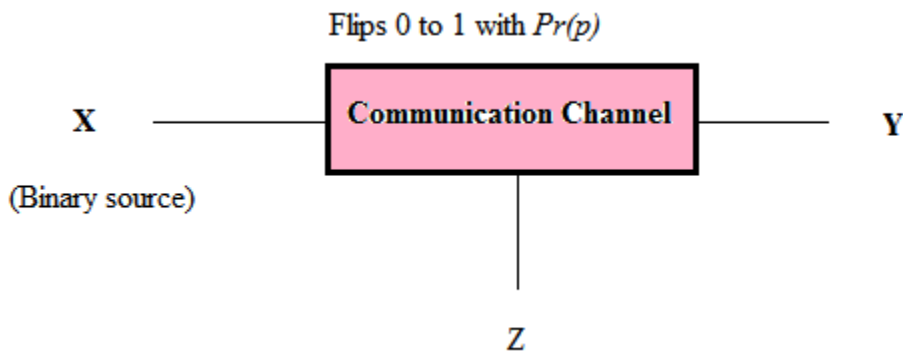


Figure 4

In general, as Figure 4 shown, suppose we have a data sources $X \in F^n = \{0, 1\}^n$ going through a communication channel, which flips 0 to 1 with probability of $\Pr(p)$, with some noises Z . the output Y of the channel is known. The process we discussed above is used to recover X by finding Z and calculating ΦY . Such processing code is called Linear Error-Correcting code.